



Digital Media Laboratory
Advanced Information & Communication Technology Center
Sharif University of Technology

Signals & Systems

Time and Frequency Characterization of Signals and Systems

Adapted from: Lecture notes from MIT and Concordia University

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Fall 2012

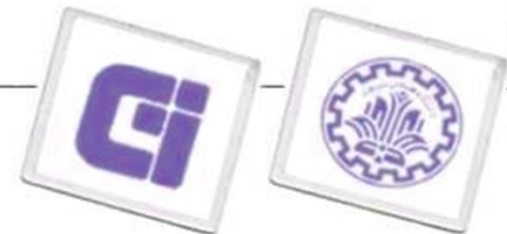
Magnitude and Phase of FT, and Parseval Relation

CT

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

$$X(j\omega) = |X(j\omega)| e^{j\angle X(j\omega)}$$

Parseval Relation $\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} \frac{1}{2\pi} |X(j\omega)|^2 d\omega$



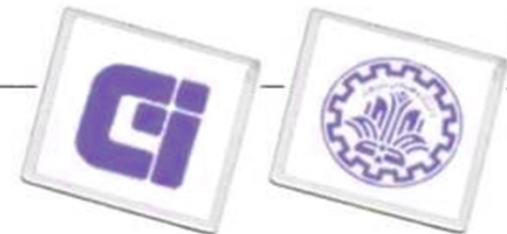
Magnitude and Phase of FT, and Parseval Relation

DT

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

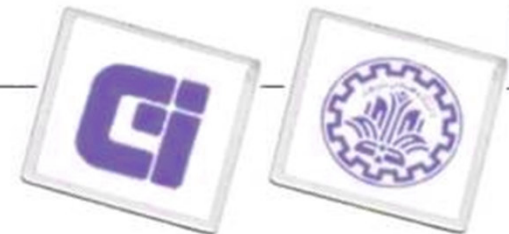
$$X(e^{j\omega}) = |X(e^{j\omega})| e^{j\angle X(e^{j\omega})}$$

Parseval Relation $\sum_{n=-\infty}^{\infty} |x[n]|^2 = \int_{2\pi} \frac{1}{2\pi} |X(e^{j\omega})|^2 d\omega$



Effects of Phase

- *Not* on signal energy distribution as a function of frequency
- Can have dramatic effect on signal shape/character
 - Constructive/Destructive interference
- Is that important?
 - Depends on the signal and the context

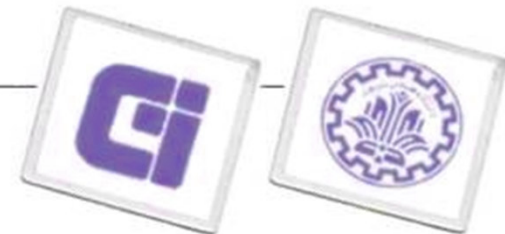


Effects of Phase

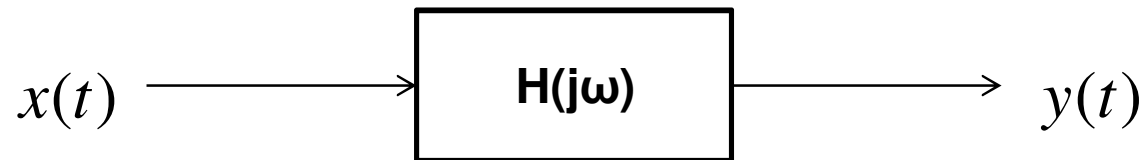
Demo:

Effect of phase on Fourier Series

Applet 17



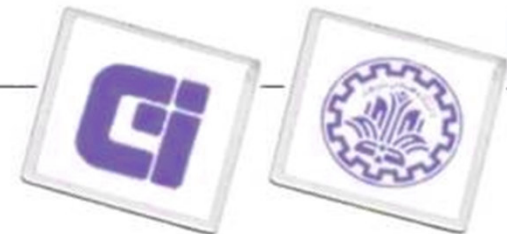
Log-Magnitude and Phase



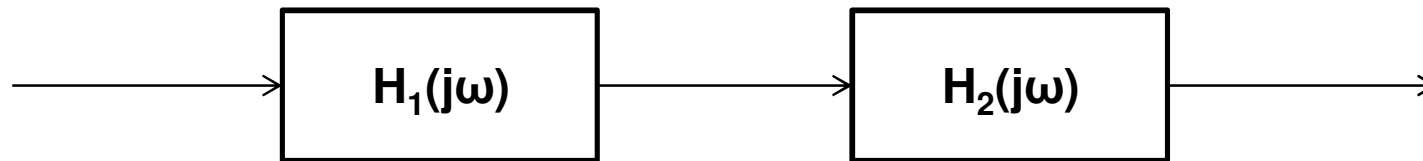
$$|Y(j\omega)| = |H(j\omega)| \cdot |X(j\omega)|$$

$$\log |Y(j\omega)| = \log |H(j\omega)| + \log |X(j\omega)|$$

$$\angle Y(j\omega) = \angle H(j\omega) + \angle X(j\omega)$$

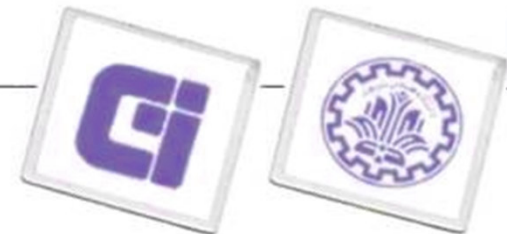


Log-Magnitude and Phase



$$\log |H(j\omega)| = \log |H_1(j\omega)| + \log |H_2(j\omega)|$$

$$\angle H(j\omega) = \angle H_1(j\omega) + \angle H_2(j\omega)$$

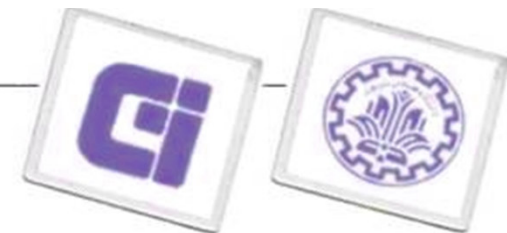
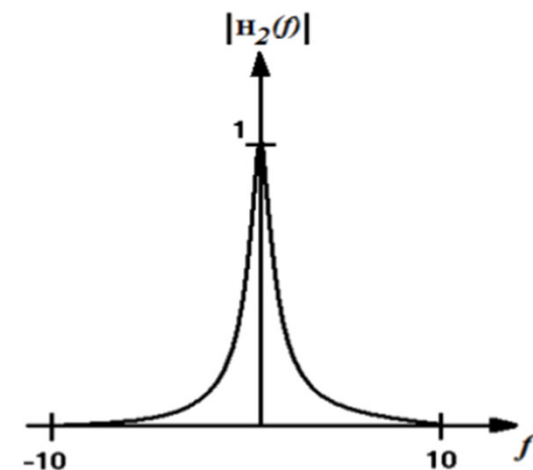
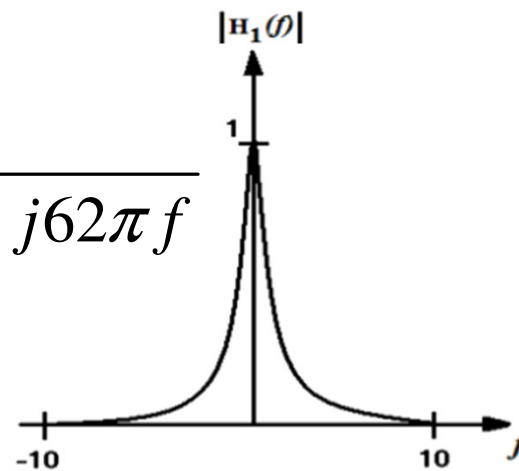


Bode Diagram

- ✧ Although linear plots, e.g., $H(j\omega)$ versus ω , of frequency responses are accurate, however, they do not always reveal important system behavior
- ✧ The plots of the two quite different-looking frequency responses look identical

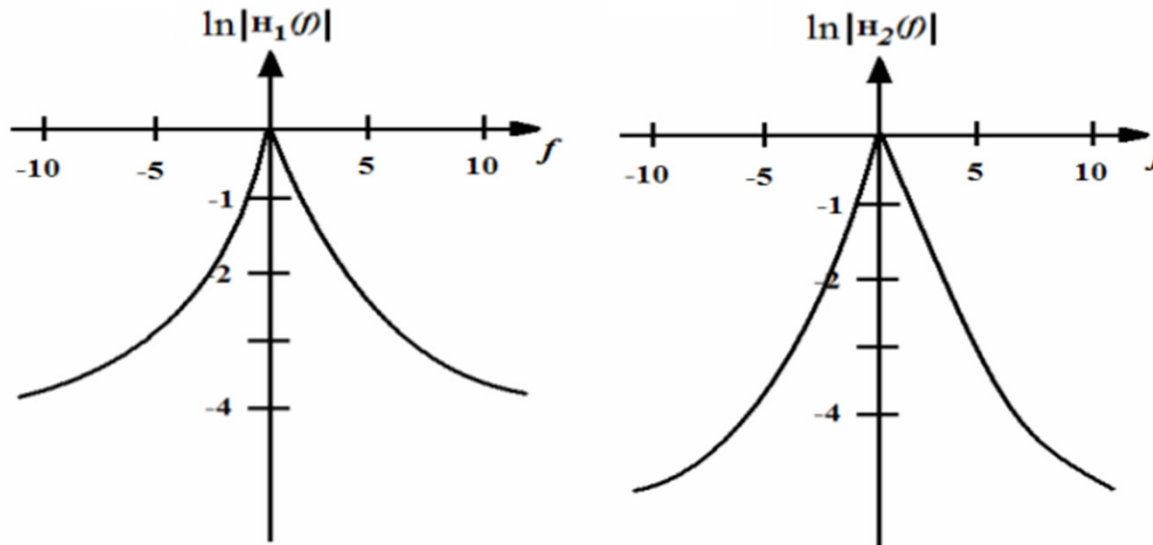
$$H_1(f) = \frac{1}{j2\pi f + 1}$$

$$H_2(f) = \frac{30}{30 - 4\pi^2 f^2 + j62\pi f}$$

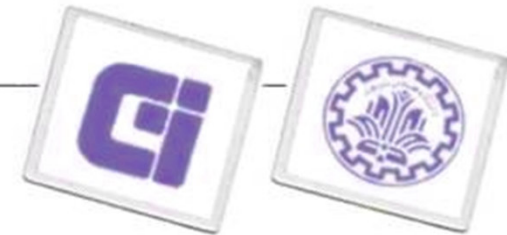


Bode Diagram

- ✧ If we plot the logarithm of magnitude instead of the magnitude, the difference can be seen more easily

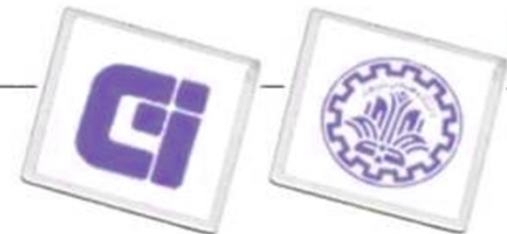


Log-magnitude plots of the two frequency responses



Bode diagram

- ✧ A more common way of displaying frequency response is the Bode diagram or Bode plot
 - ✧ A log-magnitude plot is logarithmic in one dimension
 - ✧ A Bode diagram is logarithmic in both dimensions
- ✧ A bode diagram is a plot of the logarithm of the magnitude of a frequency response against a logarithmic frequency scale
 - ✧ $\log |H(j\omega)|$ versus scaled ω
- ✧ In a bode diagram, $|H(j\omega)|$ is converted to a logarithmic scale using the unit decibel (dB)



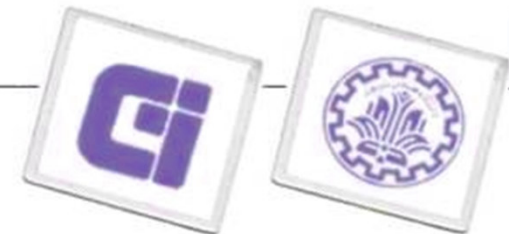
Plotting Log-Magnitude and Phase

- For real-valued signals and systems

$$\left. \begin{aligned} |H(-j\omega)| &= |H(j\omega)| \\ \angle H(-j\omega) &= -\angle H(j\omega) \end{aligned} \right\} \Rightarrow \text{Plot for } \omega \geq 0, \text{ often with a } \mathbf{logarithmic} \text{ scale for frequency in CT}$$

- In DT, need only plot for $0 \leq \omega \leq \pi$ (with *linear* scale)
- For historical reasons, log-magnitude is usually plotted in units of *decibels* (dB):

$$1 \text{ bel} = 10 \text{ decibels} = \frac{\text{output power}}{\text{input power}} = 10$$



Plotting Log-Magnitude and Phase

$$10 \log_{10} |H(j\omega)|^2 \stackrel{\text{power}}{=} 20 \log_{10} |H(j\omega)| \stackrel{\text{magnitude}}{=}$$

$$|H(j\omega)| = 1 \quad \rightarrow \quad 0dB$$

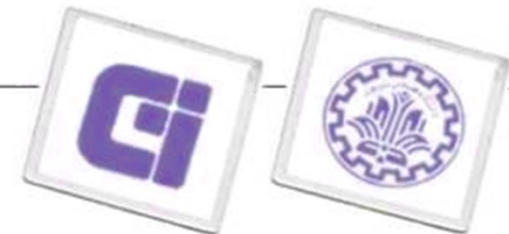
$$|H(j\omega)| = \sqrt{2} \quad \rightarrow \quad \sim 3dB$$

$$|H(j\omega)| = 2 \quad \rightarrow \quad \sim 6dB$$

$$|H(j\omega)| = 10 \quad \rightarrow \quad 20dB$$

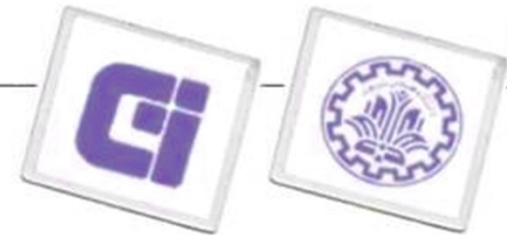
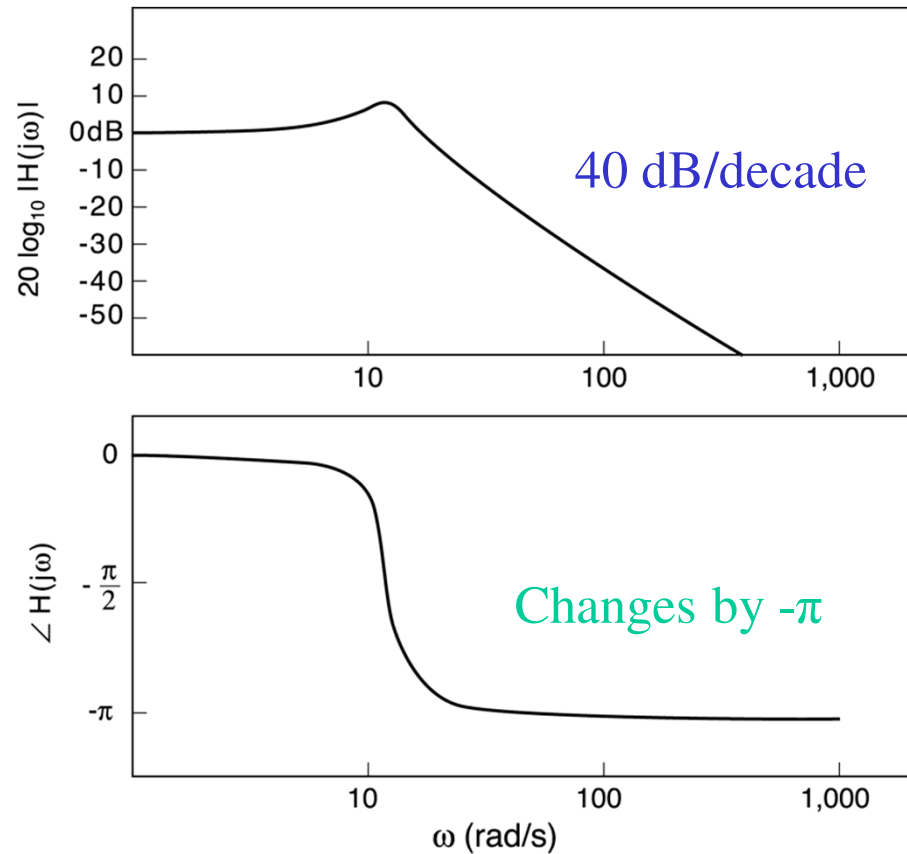
$$|H(j\omega)| = 100 \quad \rightarrow \quad 40dB$$

So... 20 dB or 2 *bels*
= 10 amplitude gain
= 100 power gain



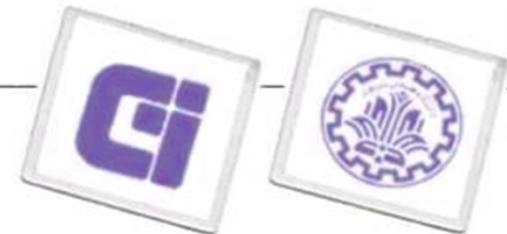
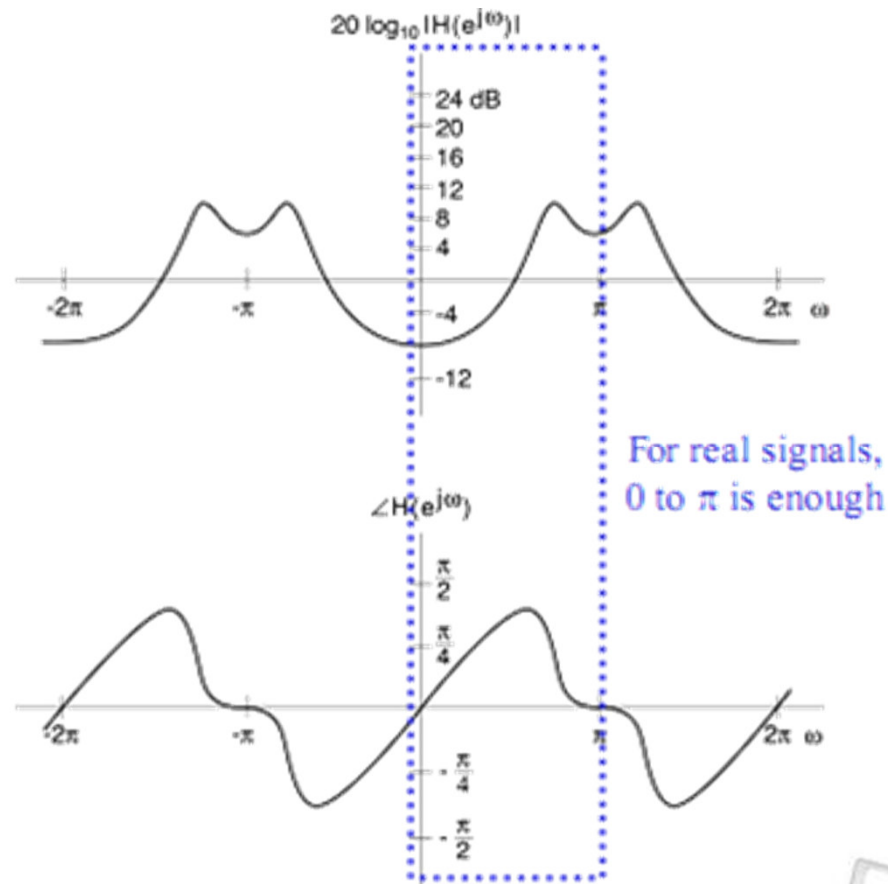
A Typical Bode plot for a second-order CT system

$20 \log |H(j\omega)|$ and $\angle H(j\omega)$ vs. $\log \omega$



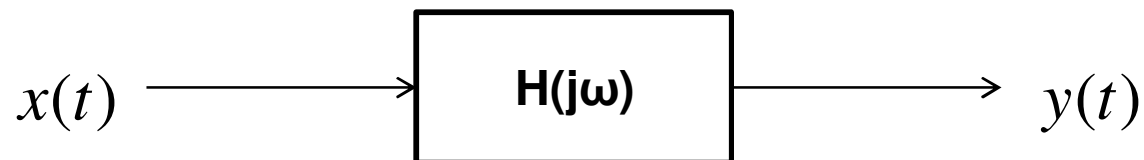
A Typical Bode plot of the magnitude and phase for a second-order DT frequency response

$20 \log |H(e^{j\omega})|$ and $\angle H(e^{j\omega})$ vs. ω



Linear Phase

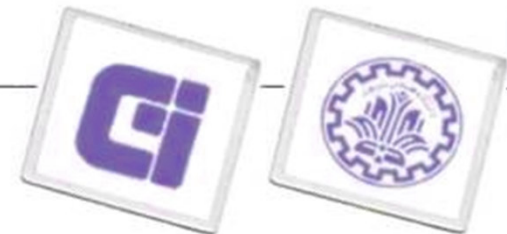
CT



$$H(j\omega) = e^{-j\omega\alpha} \Rightarrow |H(j\omega)| = 1, \angle H(j\omega) = -\alpha\omega \text{ (Linear in } \omega)$$

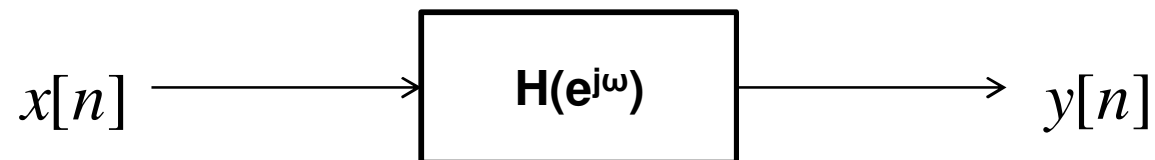
$$Y(j\omega) = e^{-j\omega\alpha} X(j\omega) \overset{\text{time-shift}}{\Leftrightarrow} y(t) = x(t - \alpha)$$

Result: Linear phase \Leftrightarrow simply a rigid shift in time, **no distortion**
 Nonlinear phase \Leftrightarrow **distortion** as well as shift



Linear Phase

DT

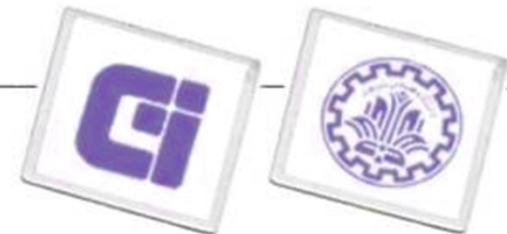


$$y[n] = x[n - n_0] \xleftrightarrow{\mathfrak{F}} Y(e^{j\omega}) = e^{-j\omega n_0} X(e^{j\omega})$$

$$H(e^{j\omega n_0}) = e^{-j\omega n_0} \Rightarrow |H(e^{j\omega})| = 1, \angle H(e^{j\omega}) = -n_0 \omega$$

Question:

What about $H(e^{j\omega}) = e^{-j\omega \alpha}$, $\alpha \neq \text{integer}$?



All-Pass Systems

CT

$$|H(j\omega)| = 1$$

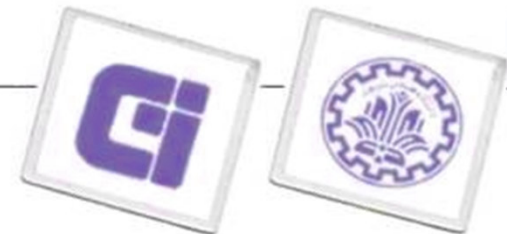
$$H(j\omega) = e^{-j\alpha\omega}$$

Linear phase

$$H(j\omega) = \frac{\alpha - j\omega}{\alpha + j\omega}$$

Nonlinear phase

$$|H(j\omega)| = \sqrt{\frac{\alpha^2 + \omega^2}{\alpha^2 + \omega^2}}$$



All-Pass Systems

DT

$$|H(e^{j\omega})| = 1$$

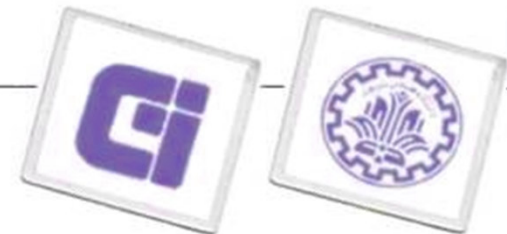
$$H(j\omega) = e^{-j\alpha\omega n_0}$$

Linear phase

$$H(j\omega) = \frac{1 - 1/2 \cdot e^{j\omega}}{1 - 1/2 \cdot e^{-j\omega}}$$

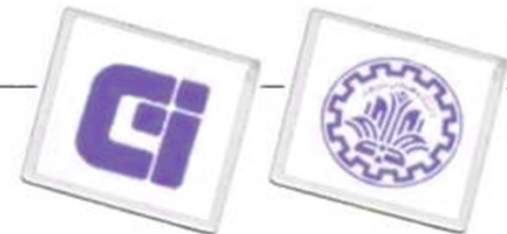
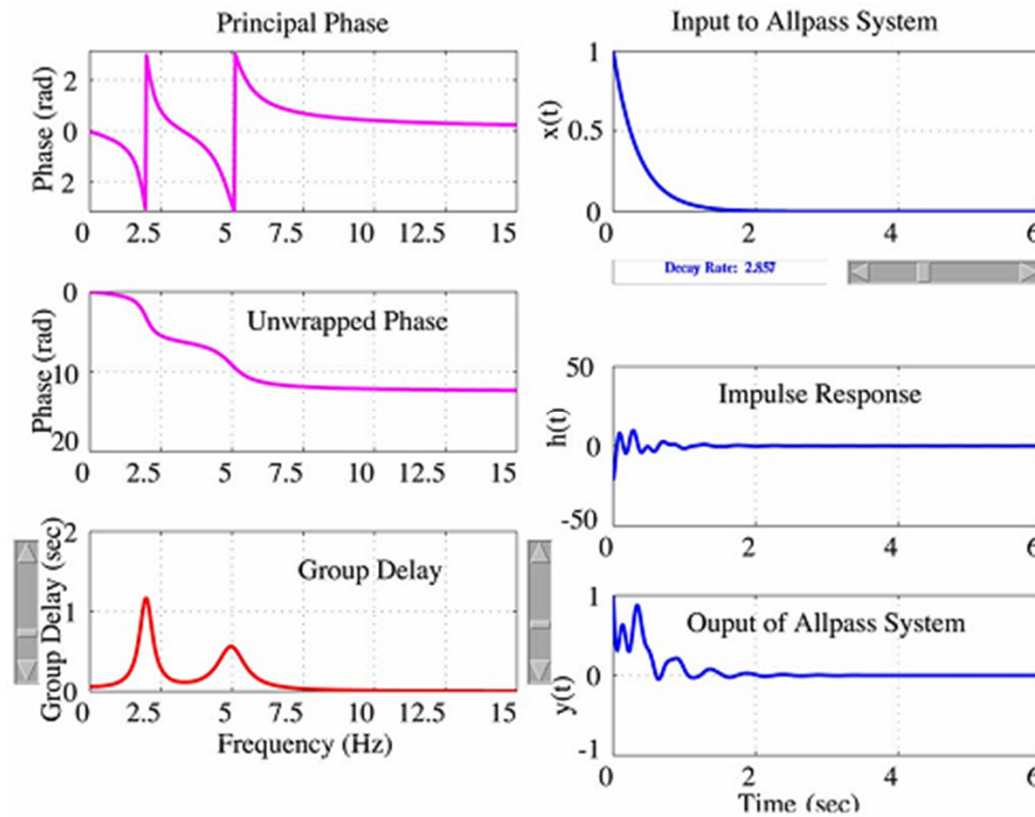
Nonlinear phase

$$|H(e^{j\omega})| = \sqrt{\frac{(1 - 1/2 \cdot \cos\omega)^2 + (1/2 \cdot \sin\omega)^2}{(1 - 1/2 \cdot \cos\omega)^2 + (1/2 \cdot \sin\omega)^2}} = 1$$



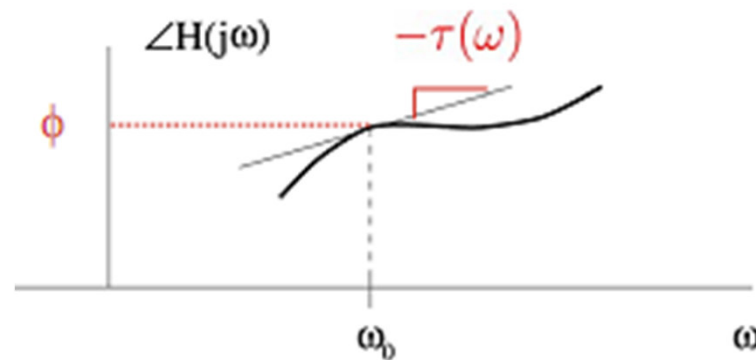
Sample

Impulse response and output of an all-pass system with nonlinear phase

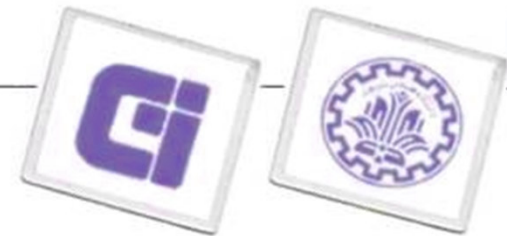


Group Delay

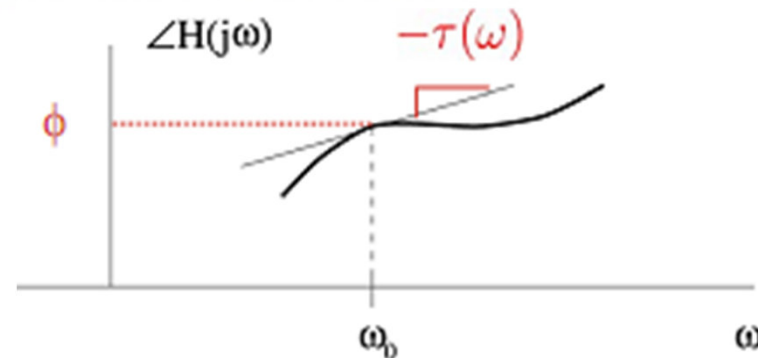
How do we think about signal delay when the phase is nonlinear?



When the signal is narrow-band and concentrated near ω_0 , $\angle H(j\omega) \sim$ linear with ω near ω_0 , then $-\frac{d\angle H(j\omega)}{d\omega}$ instead of $\frac{\angle H(j\omega)}{d\omega}$ reflects the time delay.



Group Delay



For frequencies "near" ω_0

$$\angle H(j\omega) \approx \angle H(j\omega_0) - \tau(\omega_0)(\omega - \omega_0) = \phi - \tau(\omega_0) \cdot \omega$$

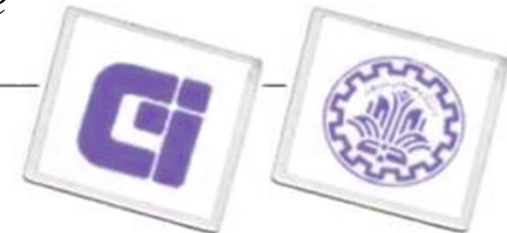
$$\tau(\omega) = -\frac{d}{d\omega} \{ \angle H(j\omega) \} = \text{Group Delay}$$

⇓

For ω near ω_0

$$H(j\omega) \approx |H(j\omega_0)| e^{j\phi} e^{-j\tau(\omega_0)\omega}$$

$$\Rightarrow e^{j\omega t} \rightarrow \sim |H(j\omega)| e^{j\phi} e^{j\omega(t - \tau(\omega_0))}$$



Linear phase / Group delay

- ✧ **Group delay** is a measure of the transit time of a signal through a system (or device) versus frequency
- ✧ **Linear phase** is a property of a filter
 - ✧ The phase response of the filter is a linear function of frequency
- ✧ Linear phase filter has the property of the true time delay
- ✧ A linear phase filter has constant group delay
 - ✧ All frequency components of a signal have equal delay times
 - ✧ There is no distortion due of select frequencies
- ✧ A filter with **non-linear** phase has a group delay that varies with frequency, resulting in phase distortion

